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Fast Approximated Power Iteration Subspace Tracking

Roland Badeau, Member, IEEE, Bertrand David, and Gaël Richard, Member, IEEE

Abstract—This paper introduces a fast implementation of the power iteration method for subspace tracking, based on an approximation less restrictive than the well known projection approximation. This algorithm, referred to as the fast API method, guarantees the orthonormality of the subspace weighting matrix at each iteration. Moreover, it outperforms many subspace trackers related to the power iteration method, such as PAST, NIC, NP3 and OPAST, while having the same computational complexity. The API method is designed for both exponential windows and sliding windows. Our numerical simulations show that sliding windows offer a faster tracking response to abrupt signal variations.

Index Terms—Subspace tracking, projection approximation, power iteration.

I. INTRODUCTION

The interest in subspace-based methods stems from the fact that they consist in splitting the observations into a set of desired and a set of disturbing components, which can be viewed in terms of signal and noise subspaces. These methods have applications in numerous domains including the fields of adaptive filtering, source localization, or parameter estimation [1]. The estimation of the signal subspace is commonly based on the traditional eigenvalue decomposition (EVD) or singular value decomposition (SVD). However, the main drawback of these decompositions is their inherent complexity. Therefore, there is a real need for fast subspace tracking techniques in the context of adaptive signal processing.

Due to this interest, a large number of approaches have already been introduced. A reference method is I. Karasalo’s algorithm [2], which involves the full SVD of a small matrix. A fast tracking method based on Givens rotations (the FST algorithm) is proposed in [3]. Other approaches consist in interlacing a recursive update of the estimated correlation matrix or the data matrix with one or a few steps of a standard SVD or power iteration algorithm. This is the case of the Jacobi SVD method [4], the transposed QR-iteration [5], the orthogonal / bi-orthogonal iteration [6], [7], and the power method [8]. Other matrix decompositions have also successfully been used in subspace tracking (for example the rank-revealing QR factorization [9], the rank-revealing URV decomposition [10], and the Lankzos (bi-)diagonalization [11]). Other techniques rely on the noise and signal subspace averaging method [12], the maximum likelihood principle [13], the operator restriction analysis [14], or the perturbation theory [15].

The estimation of the signal subspace can also be viewed as a constrained or unconstrained optimization problem [16]–[21], for which the introduction of a projection approximation hypothesis lead to fast subspace tracking methods (see e.g. the PAST [22] and NIC [23] algorithms). In [8], it is shown that these subspace trackers are closely linked to the classical power iteration method [24]. Several implementations of this method based on QR factorizations are proposed in [6], among which the Loraf2 and Loraf3 algorithms. However, compared to PAST and NIC, Loraf2 is more computationally demanding and the performance of Loraf3 is degraded. Another fast implementation of the power iteration method, the NP3 algorithm which relies on rank-one matrix updates, is proposed in [8], but our numerical simulations showed that this algorithm does not converge in many situations. An orthonormal version of the PAST algorithm, proposed in [25], can be seen as a fast implementation of the power method and outperforms PAST, NIC and NP3. Concurrently, the recent API method [26], based on the power iteration method and on a new projection approximation, has the same computational complexity as the above mentioned algorithms but provides a better estimation of the dominant subspace.

All these adaptive techniques are designed for exponential windows. Indeed, this choice tends to smooth the variations of the signal parameters, and thus allows a low-complexity update at each time step. However, it is only suitable for slowly changing signals. Conversely, a few subspace trackers are based on sliding windows, which generally require more computations, but offer a faster tracking response to sudden signal changes [22], [27]–[30]. In particular, a sliding window version of the API algorithm is proposed in [31].

This paper presents several fast implementations of the API method. These algorithms present several advantages:

- they can be applied either on an infinite exponential window or on a truncated window (e.g. a sliding window which may have an exponential decrease),
- an orthonormal subspace basis is computed at each time step, which is required for some subspace-based estimation methods, such as MUSIC [32],
- they rely on a new projection approximation, less restrictive than the classical one, which leads to better tracking results. In particular, it is shown that the PAST and OPAST subspace trackers can be viewed as approximations of the fast API method.

The paper is organized as follows. In section II, we introduce the various window shapes applied to the data. In section III, the classical power iteration method is reviewed, then the projection approximation is discussed in section IV.

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Our approximated power iteration (API) method is introduced in section V, and a fast implementation of this algorithm is proposed in section VI. In section VII, it is shown that both PAST and OPAST can be seen as approximations of the fast API algorithm. In section VIII, the performance of this method is compared to that of several subspace trackers, among which PAST and OPAST. Finally, the main conclusions of this paper are summarized in section IX.

II. DATA WINDOWING

Let \( \{ x(t) \}_{t \in \mathbb{Z}} \) be a sequence of \( n \)-dimensional data vectors. We are interested in computing the dominant subspace spanned by its correlation matrix. This matrix can be estimated according to the nature of the data window.

A. Exponential window

The estimated \( n \times n \) correlation matrix is defined as

\[
C_{xx}(t) = \sum_{u=-\infty}^{t} \beta^{t-u} x(u) x(u)^H
\]

where \( 0 < \beta < 1 \) is the forgetting factor. It can be recursively updated according to the following scheme:

\[
C_{xx}(t) = \beta C_{xx}(t-1) + x(t) x(t)^H.
\] (1)

B. Truncated window

The \( n \times n \) correlation matrix \( C_{xx}(t) \) is estimated on a window of length \( l \in \mathbb{N}^+ \):

\[
C_{xx}(t) = \sum_{u=t-l+1}^{t} \beta^{t-u} x(u) x(u)^H.
\] (2)

where \( 0 < \beta \leq 1 \). The case \( \beta = 1 \) corresponds to a rectangular (or sliding) window. This matrix can be recursively updated according to the following scheme:

\[
C_{xx}(t) = \beta C_{xx}(t-1) + x(t) x(t)^H - \beta x(t-l) x(t-l)^H.
\] (3)

C. Unified formalization

Both equations (1) and (3) can be written in the form

\[
C_{xx}(t) = \beta C_{xx}(t-1) + \bar{x}(t) \bar{x}(t)^H
\] (4)

where \( \bar{x}(t) \) and \( \bar{J} \) are defined according to the window shape:

- in the exponential window case:

\[
\bar{J} = 1
\] (5)

\[
\bar{x}(t) = x(t)
\] (6)

- in the truncated window case:

\[
\bar{J} = \begin{bmatrix} 1 & 0 \\ 0 & -\beta \end{bmatrix}
\] (7)

\[
\bar{x}(t) = \begin{bmatrix} x(t) \\ x(t-l) \end{bmatrix}
\] (8)

Let \( p \) be the rank of the update involved in equation (4). Since \( p = 1 \) in the exponential window case and \( p = 2 \) in the truncated window case, \( p \) characterizes the window shape. In particular, \( \bar{x}(t) \) is a \( n \times p \) matrix and \( \bar{J} \) is a \( p \times p \) matrix.

III. THE CLASSICAL POWER ITERATION METHOD

The power iteration method [8] tracks the dominant subspace\(^1 \) of dimension \( r \leq n \) spanned by the \( n \times n \) matrix \( C_{xx}(t) \). At each time step, a basis of this subspace is computed, represented by an orthonormal matrix \( W(t) \) of dimension \( n \times r \). The computation of \( W(t) \) consists of a data compression step (9) and an orthonormalization step (10) of the compressed matrix at each iteration:

\[
C_{xy}(t) = C_{xx}(t) W(t-1)
\] (9)

\[
W(t) R(t) = C_{xy}(t).
\] (10)

where \( C_{xy}(t) \) can be considered as a \( n \times r \) correlation matrix between the \( n \)-dimensional data vectors \( x(t) \) and the \( r \)-dimensional compressed data vectors

\[
y(t) = W(t-1)^H x(t).
\] (11)

The orthonormalization step (10) involves a \( r \times r \) matrix \( R(t) \), such that \( R(t)^H R(t) = \Phi(t) \), where \( \Phi(t) \) is the \( r \times r \) positive definite matrix \( C_{xx}(t)^H C_{xy}(t) \). Consequently, \( R(t)^H \) is a square root of \( \Phi(t) \). In particular, \( R(t)^H \) is equal to the positive definite square root of \( \Phi(t) \), right multiplied by a unitary matrix\(^2 \). For example, \( R(t) \) can be triangular [6], or positive definite [8].

If \( C_{xx}(t) \) remains constant and if its first \( r \) eigenvalues are strictly larger than the \( (n-r)^{th} \) others, the power iteration method converges globally and exponentially to the principal subspace [8] [24, pp. 410-411]. Note that the multiplication in step (9) involves \( n^2 r \) operations, and the orthonormalization step (10) requires \( O(n^2 r^2) \) operations\(^3 \). Because of its high computational cost, this algorithm is not suitable for real-time processing.

IV. THE PROJECTION APPROXIMATION

We are now looking for an approximation that will allow us to reduce the complexity. Suppose that \( W(t-1) \) exactly spans the \( r \)-dimensional dominant subspace of \( C_{xx}(t) \). Then equation (9) yields

\[
C_{xy}(t) = W(t-1) C_{yy}(t)
\] (12)

where the matrix \( C_{yy}(t) \triangleq W(t-1)^H C_{xx}(t) W(t-1) \) can be seen as the correlation matrix of the compressed data

\(^1\)The \( r \)-dimensional dominant subspace of the positive semidefinite matrix \( C_{xx}(t) \) is the subspace spanned by the \( r \) eigenvectors of \( C_{xx}(t) \) associated to the \( r \) eigenvalues of highest magnitude (which are supposed to be strictly greater than the \( n-r \) others).

\(^2\)If \( T \) is a positive definite matrix, a square root of \( T \) is any matrix \( S \) of the same dimension such that \( S^2 = T \). Such a matrix is denoted \( S = T^{1/2} \). There is only one positive definite square root of \( T \). The other square roots are obtained by right multiplying this positive definite square root by any unitary transformation. The notation \( S^{\pm 2} \) can denote any of them.

\(^3\)In this paper, operations counts are expressed in terms of multiply / accumulate (MAC) operations, herein referred to as flops. Whenever a specific matrix function is used, such as orthonormalization, inversion or square rooting, only the order of the operations count is presented, since the exact operations count depends on the way this function is implemented. Nevertheless, \( r \) is supposed to be much lower than \( n \), so that the dominant cost of the power iteration method is that of the first step, whose exact operations count is known \( (n^2 r) \).
In this case, $W(t)$ and $W(t-1)$ are two orthonormal matrices spanning the range space of $C_{xy}(t)$, thus

$$W(t) = W(t-1) \Theta(t)$$

(13)

where $\Theta(t) \triangleq W(t-1)^H W(t)$ is a $r \times r$ orthonormal matrix. Substituting equation (12) into equation (10) and left multiplying by $W(t)^H$ yields the polar decomposition of $R(t)^H$:

$$R(t)^H = C_{yy}(t) \Theta(t)$$

(14)

where $C_{yy}(t)$ is the positive definite factor and $\Theta(t)$ is the orthonormal factor. Now suppose that $W(t-1)$ approximately spans the dominant subspace of $C_{xx}(t)$. Then equations (13) and (14) become approximations:

$$W(t) \simeq W(t-1) \Theta(t)$$

(15)

$$R(t)^H \simeq C_{yy}(t) \Theta(t)$$

(16)

where the $r \times r$ matrix $\Theta(t)$ is nearly orthonormal.

Compared to equation (15), the classical projection approximation [22] is equivalent to $W(t) \simeq W(t-1)$ at each time step. The validity of this approximation additionally requires that $\Theta(t)$ is close to the $r \times r$ identity matrix (herein denoted $I_r$). In this case, equation (16) shows that $R(t)^H$ must be nearly positive definite. Consequently, the choice of the square root $R(t)^H$ of $\Phi(t)$ is restricted (e.g. $R(t)$ can no longer be upper triangular, as it was in [6]).

The NP3 implementation of the power method [8] is based on this approximation, but this algorithm relies on a matrix $R(t)$ which deviates from the positive definite structure constraint. Therefore, the classical projection approximation does not stand, and this subspace tracker is not guaranteed to converge.

Concurrently, the algorithms presented in section V do not have to face this limitation, since they rely on the less restrictive approximation (15). Also note that (15) is the best approximation of $W(t)$ in terms of mean square error, since the solution to the minimization problem

$$\arg \min_{\Theta \in C^{r \times r}} \| W(t) - W(t-1) \Theta \|^2_F$$

is $\Theta(t) = W(t-1)^H W(t)$ (where $W(t-1)$ is supposed to be orthonormal).

### V. Approximated Power Iteration

The complexity of the power iteration method can be reduced by introducing approximation (15) at time $t - 1$ in step (9). Then the $n \times r$ matrix $C_{xy}(t)$ can be updated recursively, as shown in section V-A, and factorization (10) can be updated, as shown in section V-C. This fast update requires the introduction of a $r \times r$ auxiliary matrix $Z(t)$, introduced in section V-B.

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3In fact, the projection approximation in [22] is defined as $W(t)^H x(t) \approx W(t-1)^H x(t)$, $\hat{y}(t) \forall t' \geq t$. It was shown in [8, pp. 301] that this approximation is equivalent to $W(t) \approx W(t-1)$ at each time step.

4Conversely, if $R(t)^H$ is chosen close to the only positive definite square root of $\Phi(t)$, the approximate polar decomposition (16) shows that $\Theta(t) \approx I_r$, so that equation (15) yields $W(t) \approx W(t-1)$.

### A. Recursion for the matrix $C_{xy}(t)$

It is shown in this section that the $n \times r$ matrix $C_{xy}(t)$ can be updated in the same way as the $n \times n$ matrix $C_{xx}(t)$ in equation (4):

$$C_{xy}(t) = \beta C_{xy}(t-1) \Theta(t-1) + \vec{z}(t) \vec{y}(t)^H$$

(17)

In the exponential window case, equation (17) involves a rank-one update ($\vec{z}(t)$ and $\vec{y}(t)$ are vectors and $J$ is a scalar), whereas in the truncated window case it involves a rank-two update ($\vec{z}(t)$ and $\vec{y}(t)$ are two-column matrices and $J$ is a $2 \times 2$ matrix).

1) **Truncated window**:

First, equation (2) can be written

$$C_{xx}(t) = X(t) D X(t)^H$$

(18)

where $X(t) \triangleq [x(t-1), x(t-2), \ldots, x(t)]$ is the $n \times l$ data matrix and $D$ is the $l \times l$ diagonal matrix $\text{diag}(\beta^1, \beta^2, \ldots, \beta, 1)$.

Substituting equation (18) into equation (9) yields

$$C_{xy}(t) = X(t) D Y(t)^H$$

(19)

where $Y(t) \triangleq W(t-1)^H X(t)$ is the $r \times l$ compressed data matrix. Now let us show recursions for matrices $X(t)$ and $Y(t)$. The first one is straightforward:

$$[x(t-l) \mid X(t)] = [X(t-1) \mid x(t)]$$

(20)

Then left multiplying equation (20) by $W(t-1)^H$ yields

$$[v(t-l) \mid Y(t)] = [W(t-1)^H X(t-1) \mid y(t)]$$

(21)

where $y(t)$, defined in equation (11), and

$$v(t-l) \triangleq W(t-1)^H x(t-l)$$

(22)

are $r$-dimensional compressed data vectors. Applying approximation (15) at time $t - 1$ to equation (21) yields the recursion

$$[v(t-l) \mid Y(t)] \simeq [\tilde{V}(t-1) \mid y(t)]$$

(23)

where $\tilde{V}(t-1)$ is the $r \times l$ compressed data matrix

$$\tilde{V}(t-1) \triangleq \Theta(t-1)^H Y(t-1)$$

(24)

From now on, the exact definition of $Y(t)$ is therefore replaced by

$$[\hat{v}(t-l) \mid Y(t)] \triangleq [\tilde{V}(t-1) \mid y(t)]$$

(25)

where the $r$-dimensional vector $\hat{v}(t-l)$, defined by the first column in the left side of equation (24), is an approximation of the vector $v(t-l)$. Equations (19), (20), (23) and (24) finally yield

$$C_{xy}(t) = \beta C_{xy}(t-1) \Theta(t-1) + \vec{x}(t) \vec{y}(t)^H - \beta^l \vec{x}(t-l) \hat{v}(t-l)^H$$

(26)

This recursion can be seen as a particular case of equation (17), where $J$ and $\vec{z}(t)$ are defined in equations (7) and (8) and the $r \times p$ (with $p = 2$) matrix

$$\hat{y}(t) \triangleq [y(t) \mid v(t-l)]$$

(27)

is an approximation of

$$y(t) \triangleq W(t-1)^H z(t) = [y(t) \mid v(t-l)]$$

(28)
2) Exponential window: substituting equation (1) into equation (9) yields
\[ C_{xy}(t) = \beta C_{xx}(t-1)W(t-1) + x(t)y(t)H. \] (28)

Applying the projection approximation (15) at time \( t-1 \), equation (28) can be replaced by the following recursion:
\[ C_{xy}(t) = \beta C_{xy}(t-1)\Theta(t-1) + x(t)y(t)H. \] (29)

This recursion can be seen as a particular case of equation (17), where \( J \) and \( x(t) \) are defined in equations (5) and (6) and the \( r \times p \) (with \( p = 1 \)) matrix \( \tilde{y}(t) \) is now equal to the vector \( y(t) \triangleq W(t-1)^{H}\tilde{x}(t) = y(t) \).

B. Recursion for the matrix \( Z(t) \)

Now, we aim at updating factorization (10) by means of equation (17). This calculation requires the introduction of an auxiliary matrix, denoted \( Z(t) \). Let \( S(t-1) \triangleq (R(t-1)\Theta(t-1))^{H} \) and suppose that the \( r \times r \) matrix \( S(t-1) \) is non-singular. Then let
\[ Z(t-1) \triangleq S(t-1)^{-1}. \] (30)

**Proposition 5.1:** The \( r \times r \) matrix
\[ S(t) \triangleq (R(t)\Theta(t))^{H} \] (31)

is non-singular if and only if the \( p \times p \) matrix \( \beta J^{-1} + y(t)^{H}h(t) \) is non-singular, where
\[ h(t) \triangleq Z(t-1)\tilde{y}(t). \] (32)

has dimension \( r \times p \). In this case, the \( r \times r \) matrix
\[ Z(t) \triangleq S(t)^{-1} \] (33)

satisfies the recursion
\[ Z(t) = \frac{1}{\beta} \Theta(t)^{H}(I_r - g(t)y(t)^{H})Z(t-1)\Theta(t)^{-H} \] (34)

where \( g(t) \) is the \( r \times p \) matrix
\[ g(t) \triangleq h(t)(\beta J^{-1} + y(t)^{H}h(t))^{-1}. \] (35)

**Proof:**
Substituting equation (10) into equation (17) and left multiplying by \( W(t-1)^{H} \) leads to
\[ \Theta(t)R(t) = \beta S(t-1)^{H} + y(t)\tilde{J}\tilde{y}(t)H. \] (36)

Next, the following matrix inversion lemma [33, pp. 18-19] will be applied to invert the right member of this equality. The interest of this approach is that the \( r \times r \) matrix inversion problem is converted into a smaller \( p \times p \) matrix inversion (with \( p = 1 \) or 2).

**Lemma 5.2:** Let \( A \) be a \( r \times r \) non-singular complex matrix. Consider the \( r \times r \) matrix \( B \triangleq A + PJQ \), where \( P, J \) and \( Q \) have dimensions \( r \times m, m \times m \) and \( m \times r \), and \( J \) is supposed to be non-singular. Then \( B \) is non-singular if and only if \( J^{-1} + QA^{-1}P \) is non-singular, and in this case
\[ B^{-1} = A^{-1} - A^{-1}P(J^{-1} + QA^{-1}P)^{-1}QA^{-1}. \] (37)

Lemma 5.2 applied to equation (36) shows that the \( r \times r \) matrix \( \Theta(t)R(t) \) is non-singular if and only if the \( p \times p \) matrix \( \beta J^{-1} + y(t)^{H}h(t) \) is non-singular (which provides a fast way of detecting the singularity of \( R(t) \) or \( \Theta(t) \)). In the non-singular case, lemma 5.2 leads to the equation
\[ (\Theta(t)R(t))^{-1} = \frac{1}{\beta} Z(t-1)^{H}(I_r - y(t)g(t)^{H}) \] (38)

Finally, left multiplying the complex conjugate transpose of this last equation by \( \Theta(t)^{H} \) and right multiplying it by \( \Theta(t)^{-H} \) yields recursion (34).

C. Recursion for the matrix \( W(t) \)

Next, proposition 5.3 introduces a fast update for the subspace weighting matrix.

**Proposition 5.3:** If the \( p \times p \) matrix \( \beta J^{-1} + y(t)^{H}h(t) \) is non-singular, \( W(t) \) satisfies the recursion
\[ W(t) = (W(t-1) + e(t)\tilde{y}(t)^{H})\Theta(t) \] (37)

where \( e(t) \) is the \( n \times p \) matrix
\[ e(t) \triangleq \tilde{x}(t) - W(t-1)y(t). \] (38)

**Proof:**
Substituting equation (10) into equation (17) and right multiplying by \( \Theta(t) \) shows that \( W(t) \) satisfies the recursion
\[ W(t)S(t)^{H} = (\beta W(t-1)S(t-1)^{H} + x(t)\tilde{J}\tilde{y}(t)^{H})\Theta(t) \] (39)

Substituting equations (36) and (38) into the above equation yields
\[ W(t)S(t)^{H} = W(t-1)\Theta(t)S(t)^{H} + e(t)\tilde{J}\tilde{y}(t)^{H}\Theta(t). \] (40)

However, left multiplying (36) by \( g(t)^{H} \) and replacing \( e(t) \) by its definition in equation (35) leads to
\[ g(t)^{H}\Theta(t)R(t) = (\beta J^{-1} + y(t)^{H}h(t))^{-H} \left[ (\beta S(t-1)h(t)^{H})^{H} + (y(t)^{H}h(t))^{H} J\tilde{y}(t)^{H} \right]. \] (41)

Then equations (32) and (30) show that
\[ (\beta S(t-1)h(t)^{H})^{H} = \beta \tilde{y}(t)^{H} = \beta J^{-1} J\tilde{y}(t)^{H}. \] (42)

Substituting equation (41) into equation (40) yields
\[ g(t)^{H}\Theta(t)R(t) = J\tilde{y}(t)^{H}. \] (43)

Finally, substituting equation (42) into equation (39) and right multiplying by \( S(t)^{H} = Z(t)^{H} \) yields equation (37).

**Note** that if \( \beta J^{-1} + y(t)^{H}h(t) \) is singular, \( Z(t) \) and \( W(t) \) can no longer be updated with equations (34) and (37). In practice, we never encountered this rank deficiency case in our numerical simulations.

A solution consists in computing \( W(t) \) and \( R(t) \) by means of a SVD or a QR factorization of \( C_{xy}(t) \). Then \( \Theta(t) = W(t-1)^{H}W(t) \) can be deduced. Note that the whole processing requires \( O(nr^{2}) \) operations; this technique must be used while \( R(t) \) or \( \Theta(t) \) remains singular. When both \( R(t) \) and \( \Theta(t) \) become non-singular again, then \( Z(t) \) can be computed, and the algorithm can switch back to the fully adaptive processing.
TABLE I
EXPONENTIAL WINDOW API ALGORITHM

<table>
<thead>
<tr>
<th>Initialization</th>
<th>( W(0) = \begin{bmatrix} I_r &amp; 0 \end{bmatrix} )</th>
<th>( Z(0) = I_r )</th>
<th>For each time step do</th>
<th>Cost</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input vector: ( x(t) )</td>
<td>( \mathbf{PAST\ main\ section} )</td>
<td></td>
<td></td>
<td>( n + O(r^2) )</td>
</tr>
<tr>
<td>( y(t) = W(t-1)^{-1} x(t) ) (11)</td>
<td>( g(t) = \frac{1}{2} \Theta(t)^H \left( I_r - g(t) y(t)^H \right) ) (34)</td>
<td>( Z(t) = (W(t-1) + e(t) g(t)^H) \Theta(t) ) (37)</td>
<td>( n r^2 + n r )</td>
<td></td>
</tr>
<tr>
<td>( W(t) = \left( W(t-1) + e(t) g(t)^H \right) \Theta(t) )</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Since \( W(t-1) \) is orthonormal, \( e(t) \) is orthogonal to \( W(t-1) \). Moreover, the orthonormality of \( W(t) \), associated to equation (37), yields
\[
\Theta(t) \Theta(t)^H = (I_r + g(t) (e(t) e(t)^H g(t))^H)^{-1}. \tag{43}
\]

Therefore, \( \Theta(t) \) is an inverse square root of the \( r \times r \) positive definite matrix \( I_r + g(t) (e(t) e(t)^H g(t))^H \). The choice of this inverse square root does not affect the subspace tracking performance.

The pseudo-code of the exponential window API algorithm is presented in Table I, and that of the truncated window API algorithm (TW-API) is presented in Table II. It can be noted that the first section of API is exactly the same as that of the PAST subspace tracker [22]; it requires only \( n r + r^2 + O(r) \) operations per time step, while the rest of the algorithm has a \( n r^2 + o(n r^2) \) computational complexity. In the same way, the first section of TW-API is similar to that of the sliding window version of PAST [29]; it requires only \( 2 n r + 2 r^2 + O(r) \), while the rest of the algorithm has a \( (n + l) r^2 + o(n r^2) \) computational complexity. Note that the implementations of API and TW-API presented in Tables I and II are of limited interest, since a number of faster subspace trackers have already been proposed in the literature, which have a \( O(n r) \) complexity (among which [3], [22], [23], [25], [29], [34] are illustrated in section VIII). A faster implementation of API and TW-API is proposed in section VI.

VI. FAST API METHOD

In this section, a fast implementation of the API method is proposed, based on a particular choice of the matrix \( \Theta(t) \). It is supposed that \( \beta I_r^{-1} + g(t) h(t)^H h(t)^{-1} \) is non-singular, so that \( \Theta(t) \) is also non-singular. Below, the \( p \times p \) identity matrix is denoted \( I_p \).

A. A particular solution to equation (43)

Let \( z(t) \) be a square root of the \( p \times p \) matrix \( e(t) e(t)^H \):
\[
z(t) z(t)^H = (e(t) e(t)^H).
\tag{45}
\]
Substituting equation (45) into equation (43) and applying the matrix inversion lemma shows that
\[
\Theta(t) \Theta(t)^H = I_r - g(t) z(t)^{-1} z(t)^H g(t) z(t)^{-1} z(t)^H g(t)^H
\tag{46}
\]
where \( g(t) \) is the \( p \times p \) positive definite matrix
\[
g(t) = I_p + z(t)^H g(t) z(t).
\tag{47}
\]
Considering equation (46), we are looking for a special solution of the form
\[
\Theta(t) = I_r - g(t) z(t)^{-1} z(t)^H g(t)^H
\tag{48}
\]
• \( W(0) \) should be a \( n \times r \) orthonormal matrix,
• \( Z(0) \) should be a \( r \times r \) positive definite matrix.
Both matrices can be calculated from an initial block of data or from arbitrary initial data. The simplest way, however, is to set \( W(0) = I_{0_{(n-r) \times r}} \) and \( Z(0) = I_r \). The choice of these initial values affects the transient behavior but not the steady state performance of the algorithm.

1Lemma 5.2 is applied with \( A = I_r, P = g(t) z(t), J = I_p \) and \( Q = z(t)^H g(t)^H \). In particular, the non-singularity of \( \Theta(t) \) is equivalent to the non-singularity of \( g(t) \).
where \( g(t) \) is a \( p \times p \) non-singular matrix. The interest of this approach is that the \( r \times r \) matrix square rooting problem in equation (46) is converted into a smaller \( p \times p \) matrix square rooting. Indeed, substituting equation (48) into equation (46) yields a sufficient condition:

\[
\hat{g}(t)^{-1} + \hat{g}(t) - \hat{g}(t)^{-1} \left( L_p - \hat{g}(t) \right) \hat{g}(t)^{-1} = \hat{g}(t)^{-1},
\]

Left multiplying the two members of this last equation by \( \hat{g}(t) \) and right multiplying them by \( \hat{g}(t)^{H} \) yields the equation:

\[
\left( \hat{g}(t) - \hat{g}(t) \right) \hat{g}(t)^{-1} \left( \hat{g}(t) - \hat{g}(t) \right)^{H} = L_p,
\]

whose solution is

\[
\hat{g}(t) = \rho(t) + \hat{g}(t)^{H} \rho(t)^{H}.
\] (49)

Even if other choices would be possible, from now on we suppose that the square root of \( \hat{g}(t) \) which is involved in the above equation is the only positive definite square root. This condition guarantees that \( \hat{g}(t) \) is positive definite, so that \( \Theta(t) \) is hermitian. Then define the \( p \times p \) positive definite matrix

\[
\hat{Z}(t) = \left[ \rho(t) \right]^{-1/2} \left( g(t) \right)^{H}.
\] (50)

Substituting equation (50) into equation (48) yields

\[
\Theta(t) = I_r - \hat{g}(t) \hat{Z}(t) \left( \hat{g}(t) \right)^{H}.
\] (51)

B. Fast implementation of the particular solution

Based on the low-rank matrix update of \( \Theta(t) \) in equation (51), it is shown below that the matrices \( \hat{Z}(t), \hat{W}(t) \) and \( \hat{V}(t) \) can also be efficiently updated. Consider the \( p \times p \) matrix

\[
\hat{y}(t) = L_p - \left( \hat{g}(t)^{H} \hat{g}(t) \right) \hat{Z}(t).
\] (52)

Since \( \Theta(t) \) is non-singular, the matrix inversion lemma shows that \( \hat{y}(t) \) is also non-singular. Then substituting equation (51) into equation (34) yields

\[
\hat{Z}(t) = \frac{1}{\beta} \left( \hat{Z}(t - 1) - \left( \hat{g}(t) \right)^{H} \left( \hat{g}(t) \right)^{H} \right) \hat{Z}(t).
\] (53)

where the \( r \times p \) matrices \( \hat{h}(t) \) and \( \hat{e}(t) \) are defined by

\[
\hat{y}(t) = \frac{1}{\beta} \left( \hat{y}(t) \right) + \hat{g}(t) \hat{Z}(t),
\] (54)

\[
\hat{h}(t) = \frac{1}{\beta} \left( \hat{Z}(t - 1) \hat{y}(t) \right),
\] (55)

\[
\hat{e}(t) = \frac{1}{\beta} \left( \hat{Z}(t - 1) \hat{g}(t) - \hat{g}(t) \hat{h}(t)^{H} \hat{g}(t) \right).
\] (56)

Then substituting equation (51) into equation (37) yields

\[
\hat{W}(t) = \hat{W}(t - 1) + \left[ \hat{e}(t) \right] \left( \hat{g}(t) \right)^{H}.
\] (57)

where \( \hat{e}(t) \) is the \( n \times p \) matrix

\[
\hat{e}(t) = \hat{e}(t) \hat{y}(t) - \hat{W}(t - 1) \hat{g}(t) \hat{Z}(t).
\] (58)

\[\text{Initialization (g.t table I)}\]

For each time step do

\[
\text{Input vector: } \hat{x}(t)
\]

\[
\text{Past main section (g.t table I)}
\]

\[
\hat{x}(t) = \left( \| \hat{x}(t) \|^{2} - \| \hat{y}(t) \| \right)^{2}
\]

\[
\hat{y}(t) = \frac{1}{r} \left( \hat{y}(t) \right) + \hat{g}(t) \hat{Z}(t)
\]

\[
\hat{h}(t) = \frac{1}{r} \left( \hat{Z}(t - 1) \hat{g}(t) \right)
\]

\[
\hat{e}(t) = \frac{1}{r} \left( \hat{Z}(t) \hat{g}(t) - \hat{g}(t) \hat{h}(t)^{H} \hat{g}(t) \right)
\]

\[
\hat{W}(t) = \hat{W}(t - 1) + \left[ \hat{e}(t) \right] \left( \hat{g}(t) \right)^{H}
\]

\[
\text{Final section (g.t table I)}
\]

\[
\hat{x}(t) = \hat{x}(t) \hat{y}(t) - \hat{W}(t - 1) \hat{g}(t) \hat{Z}(t)
\]

\[
\mathbf{TABLE III}
\]

\[
\text{EXPERIMENTAL window fast API (FAPI) algorithm}
\]

\[
\begin{array}{c|c}
\text{Initialization} & \text{Cost} \\
\hline
\text{Input vector: } \hat{x}(t) & n + r \\
\text{Past main section} & \text{Cost} \\
\hat{x}(t) = \left( \| \hat{x}(t) \|^{2} - \| \hat{y}(t) \| \right)^{2} & r \\
\hat{y}(t) = \frac{1}{r} \left( \hat{y}(t) \right) + \hat{g}(t) \hat{Z}(t) & r \\
\hat{h}(t) = \frac{1}{r} \left( \hat{Z}(t - 1) \hat{g}(t) \right) & r^{2} \\
\hat{e}(t) = \frac{1}{r} \left( \hat{Z}(t) \hat{g}(t) - \hat{g}(t) \hat{h}(t)^{H} \hat{g}(t) \right) & r^{2} \\
\hat{W}(t) = \hat{W}(t - 1) + \left[ \hat{e}(t) \right] \left( \hat{g}(t) \right)^{H} & r^{2} + 3r \\
\end{array}
\]

\[
\text{VII. LINK with the PAST and OPAST algorithms}
\]

In this section, it is shown that the classical exponential window PAST algorithm can be seen as a first order approximation of the FAPI algorithm. Indeed, the error \( e(t) \) is the component of \( \hat{x}(t) \) that does not belong to the signal subspace spanned by \( \hat{W}(t - 1) \). Thus, if this subspace slowly varies upon time, and if the Signal to Noise Ratio (SNR) is high, \( e(t) \approx 0 \). If the second order term \( ||e(t)||^{2} \) is disregarded in

\[\text{Note that this implementation of FAPI is faster than that proposed in [26], whose global cost was } n(3r + 2) + 5r^{2} + O(r).\]
A. Comparison of FAPI and TW-FAPI with other existing subspace trackers

In this section, the test signal is a sum of $r = 4$ complex sinusoidal sources plus a complex white Gaussian noise (the SNR is 5.7 dB). The frequencies of the sinusoids vary according to a jump scenario originally proposed by P. Strohbehn in the context of Direction Of Arrival estimation [36]: their values abruptly change at different time instants, between which they remain constant. Their variations are represented on Figure 1-a. This signal is processed in section VIII-A.1 by means of an exponential window whose forgetting factor is $\beta \approx 0.99$, and in section VIII-A.2 by means of a sliding window of length $l = 120$. These parameters were chosen so that the effective window length is the same in both cases, i.e. $\beta = \frac{1}{1-T/T}$. Section VIII-A.3 focuses on the orthonormality of the subspace weighting matrix. The complexities of the various subspace trackers illustrated in this section are given in Table V.

<table>
<thead>
<tr>
<th>Algorithm</th>
<th>Complexity (flops)</th>
<th>Window</th>
<th>Figure</th>
</tr>
</thead>
<tbody>
<tr>
<td>FAPI</td>
<td>$n(3r+2) + 5r^2 + O(r)$</td>
<td>exponential</td>
<td>Fig. 1</td>
</tr>
<tr>
<td>PAST</td>
<td>$3nr + 2r^2 + O(r)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>NIC</td>
<td>$4nr + 2r^2 + O(r)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>OPAST</td>
<td>$n(4r+1) + 2r^2 + O(r)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Karasalo</td>
<td>$nr^2 + n(3r+2) + O(r^2)$</td>
<td>exponential</td>
<td>Fig. 2</td>
</tr>
<tr>
<td>FST</td>
<td>$n(6r+2) + 12r^2 + O(r)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Householder PAST</td>
<td>$n(4r+1) + 2r^2 + O(r)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Lora2</td>
<td>$nr^2 + n(3r+2) + O(r^2)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>SPI</td>
<td>$4nr + n(4r+2) + O(r^2)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>TW-FAPI</td>
<td>$n(6r+8) + 4r^2 + O(r^2)$</td>
<td>sliding</td>
<td>Fig. 3</td>
</tr>
<tr>
<td>SW-PAST</td>
<td>$5nr + 4r^2 + O(r)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>SW-NIC</td>
<td>$6nr + 4r^2 + O(r)$</td>
<td></td>
<td></td>
</tr>
<tr>
<td>SW-OPAST</td>
<td>$n(15r+28) + 12r^2 + O(r)$</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

TABLE V

COMPARISON OF THE COMPLEXITIES

1) Exponential window case: figure 1-b shows the maximum principal angle error trajectory $\theta_{\text{FAPI}}(t)$, obtained with the FAPI method with parameters $n = 80$ and $\beta \approx 0.99$. Then this result is compared to that obtained with the PAST subspace tracker: figure 1-c shows the ratio in dB of the trajectories obtained with FAPI and PAST, i.e.

$$20 \log_{10} \left( \frac{\theta_{\text{FAPI}}(t)}{\theta_{\text{PAST}}(t)} \right).$$

At initialization, it can be noticed that FAPI converges faster than PAST. Moreover, PAST does not provide an orthonormal subspace weighting matrix. Figure 1-d shows the ratio in dB of the trajectories obtained with FAPI and the NIC subspace tracker\(^{15}\), which is a robust generalization of PAST [23]. It can be seen that the subspace estimation error is always smaller with FAPI. As PAST, NIC does not guarantee the orthonormality of the subspace weighting matrix. Figure 1-e shows the ratio of the trajectories obtained with FAPI and OPAST. The two algorithms reach the same performance, except at initialization, where FAPI converges faster. In fact, the difference is much more distinct with the sliding window versions of these algorithms (see section VIII-A.2).

\(^{15}\)The learning step $\eta$ is equal to 0.7.

VIII. SIMULATION RESULTS

In this section, the performance of the subspace estimation is analyzed in the context of frequency estimation, in terms of the maximum principal angle between the true dominant subspace of the correlation matrix $C_{xx}(t)$ (obtained via an exact eigenvalue decomposition), and the estimated dominant subspace of the same correlation matrix (obtained with the subspace tracker). This error criterion was initially proposed by P. Comon and G.H. Golub as a measure of the distance between equidimensional subspaces [24, pp. 603-604]. In section VIII-A, the FAPI and TW-FAPI algorithms are compared to other existing subspace trackers. In section VIII-B, the behavior of the API method regarding the SNR and the parameters $n$ and $r$ is investigated.
In figure 2, the FAPI algorithm is compared to five other well-known subspace trackers:

- I. Karasalo’s algorithm [2],
- the Fast Subspace Tracking (FST) algorithm [3],
- the novel PAST algorithm employing Householder transformations, herein called Householder PAST [34],
- the Low-Rank Adaptive Filter (Loraf2) algorithm [7],
- and the Subspace Projection (SP1) algorithm [37].

Figure 2-a shows that the behaviors of FAPI and Karasalo’s algorithm are very similar. However the dominant cost of the latter is \( nr^2 \) (see table V). Figure 2-b shows that FAPI converges to the signal subspace much more precisely than FST. Moreover, FST is more computationally demanding than FAPI. Figure 2-c shows that FAPI and Householder PAST reach the same performance, except at initialization, where FAPI converges faster. Figure 2-d shows that the same remark can be made about FAPI and Loraf2. Besides, the dominant complexity of Loraf2 is \( nr^2 \).

Among the various subspace trackers that we have tested, SP1 is the only one which really outperformed FAPI (see figure 2-e). However, table V shows that SP1 is the most computationally demanding algorithm. In other respects, it is only suitable for time series data analysis, and was only designed for exponential windows.

2) Sliding window case: figure 3-a shows the maximum principal angle error trajectory \( \theta_{TW-FAPI}(t) \), obtained with the TW-FAPI method with parameters \( \beta = 1 \) (which turns the truncated window into a sliding window), \( n = 80 \) and \( l = 120 \). It can be noticed that this algorithm has a fast convergence rate after each frequency jump. This result can be compared to that of figure 1-b, obtained with the exponential window FAPI method, for which the response to frequency jumps is slower, because of the nature of the window which tends to smooth the signal variations. Figure 3-b shows the ratio in dB of the trajectories obtained with TW-FAPI and the sliding window version of PAST, herein called SW-PAST [22],[29]. It can be seen that TW-FAPI converges faster than SW-PAST at initialization. Note that as PAST, SW-PAST does not provide an orthonormal subspace weighting matrix. Figure 3-c shows the ratio in dB of the trajectories obtained with TW-FAPI and Loraf2. Figure 3-d shows the ratio in dB of the trajectories obtained with TW-FAPI and Loraf2 and SW-OPAST in regions where the frequencies are constant.

3) Orthonormality error: the orthonormality of the subspace weighting matrix \( W(t) \) can be measured by means of the following error criterion:

\[
20 \log_{10} (\| W(t)H W(t) - I_r \|_F ) .
\]

SW-NIC is also implemented with \( \eta = 0.7 \).
B. Behavior of the API method regarding the SNR and the parameters $n$ and $r$

In this section, the test signal is still a sum of $r = 4$ complex sinusoidal sources plus a complex white gaussian noise. However, the frequencies of the sinusoids are constant, equal to the initial values given in figure 1-a.

1) Influence of the SNR: in this section, the effect of the SNR onto the subspace estimation is investigated. To this end, the noise part of the test signal was synthesized so that the SNR varies linearly from +30 dB to -30 dB (see figure 4-a).

Figure 4-b shows the maximum principal angle error trajectory obtained with the FAPI method with parameters $n = 80$ and $\beta \approx 0.99$. It can be seen that the performance of the subspace estimation collapses beyond $n \approx 2600$. Figure 4-a shows that from this time instant the SNR is lower than $-10$ dB. Figure 4-c shows the maximum principal angle error trajectory obtained with the TW-FAPI method with parameters $\beta = 1$, $n = 80$ and $l = 120$. Again, the performance of the subspace estimation collapses beyond $n \approx 2600$. Although they are not illustrated here, we observed that the performance of all the above mentioned subspace trackers similarly collapse beyond the same SNR limit ($-10$ dB).

2) Influence of the ratio $n/r$: in this section, we focus on the influence of the ratio $n/r$ onto the subspace estimation. The SNR is constant, equal to $5.7$ dB.

Figure 5-a shows the mean of $\theta_{\text{FAPI}}(t)$, as a function of the ratio $n/r$, for all $n \in \{r+1, \ldots, 80\}$ (with $\beta \approx 0.99$). It can be seen that the subspace estimation becomes reliable as soon as $n/r \geq 7$. Figure 5-b shows the mean of $\theta_{\text{TW-FAPI}}(t)$, as a function of the ratio $n/r$, for all $n \in \{r+1, \ldots, 80\}$ (with $\beta = 1$ and $l = 120$). Again, it can be seen that the subspace estimation becomes reliable as soon as $n/r \geq 7$. Although they are not illustrated here, we observed that the same remark is valid for all the above mentioned subspace trackers.

3) Tracking a subspace of wrong dimension: since the dimension $r$ of the signal subspace is unknown in many applications, we investigate in this section the performance of the FAPI and TW-FAPI algorithms when applied with a wrong subspace dimension $r$. The SNR is constant, equal to $5.7$ dB. The performance of the subspace estimation is analyzed in terms of the maximum principal angle between the true $4$-dimensional signal subspace and the estimated $r$-dimensional subspace.

Figure 5-c shows the mean of $\theta_{\text{FAPI}}(t)$, as a function of $r$, for all $r \in \{1, \ldots, 20\}$ (with parameters $\beta \approx 0.99$ and $n = 80$). Similarly, figure 5-d shows the mean of $\theta_{\text{TW-FAPI}}(t)$, as a function of $r$, for all $r \in \{1, \ldots, 20\}$ (with parameters $l = 120$ and $n \approx 80$). It can be seen that the subspace estimation is reliable in all cases:

- if $r = 4$, the maximum principal angle is very low (as expected),
- if $r < 4$, the maximum principal angle remains low, which means that the estimated lower-dimensional subspace is nearly included in the true signal subspace,
our adaptive version of the ESPRIT algorithm [38].

Finally, these subspace tracking algorithms were presented, based either on exponential windows or on truncated windows. These algorithms reach a linear complexity and guarantee the orthonormality of the subspace weighting matrix at each time step. In the context of frequency estimation, the method proves able to track abrupt frequency variations robustly, and outperforms many subspace trackers, both in terms of subspace estimation and computational complexity. Finally, these subspace tracking algorithms can be considered as the starting point of a real-time frequency tracker, whose full implementation can involve our adaptive version of the ESPRIT algorithm [38].

**IX. CONCLUSIONS**

In this paper, several implementations of the API algorithm for subspace tracking were presented, either on exponential windows or on truncated windows. These algorithms reach a linear complexity and guarantee the orthonormality of the subspace weighting matrix at each time step. In the context of frequency estimation, the method proves able to track abrupt frequency variations robustly, and outperforms many subspace trackers, both in terms of subspace estimation and computational complexity. Finally, these subspace tracking algorithms can be considered as the starting point of a real-time frequency tracker, whose full implementation can involve our adaptive version of the ESPRIT algorithm [38].

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